



**SNS COLLEGE OF TECHNOLOGY**

(An Autonomous Institution)

Coimbatore-35



**16EC201 – SIGNALS & SYSTEMS**

**DISCRETE FOURIER SERIES**



# Discrete Fourier Series



- Given a periodic sequence  $\tilde{x}[n]$  with period  $N$  so that

$$\tilde{x}[n] = \tilde{x}[n + rN]$$

- The Fourier series representation can be written as

$$\tilde{x}[n] = \frac{1}{N} \sum_k \tilde{X}[k] e^{j(2\pi/N)kn}$$

- The Fourier series representation of continuous-time periodic signals require infinite many complex exponentials
- Not that for discrete-time periodic signals we have

$$e^{j(2\pi/N)(k+mN)n} = e^{j(2\pi/N)kn} e^{j(2\pi m)n} = e^{j(2\pi/N)kn}$$

- Due to the periodicity of the complex exponential we only need  $N$  exponentials for discrete time Fourier series

$$\tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] e^{j(2\pi/N)kn}$$



# Discrete Fourier Series Pair



- A periodic sequence in terms of Fourier series coefficients

$$\tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] e^{j(2\pi/N)kn}$$

- The Fourier series coefficients can be obtained via

$$\tilde{X}[k] = \sum_{n=0}^{N-1} \tilde{x}[n] e^{-j(2\pi/N)kn}$$

- For convenience we sometimes use

$$W_N = e^{-j(2\pi/N)}$$

- Analysis equation

$$\tilde{X}[k] = \sum_{n=0}^{N-1} \tilde{x}[n] W_N^{kn}$$

- Synthesis equation

$$\tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] W_N^{-kn}$$



## Example 1



- DFS of a periodic impulse train

$$\tilde{x}[n] = \sum_{r=-\infty}^{\infty} \delta[n - rN] = \begin{cases} 1 & n = rN \\ 0 & \text{else} \end{cases}$$

- Since the period of the signal is N

$$\tilde{X}[k] = \sum_{n=0}^{N-1} \tilde{x}[n] e^{-j(2\pi/N)kn} = \sum_{n=0}^{N-1} \delta[n] e^{-j(2\pi/N)kn} = e^{-j(2\pi/N)k0} = 1$$

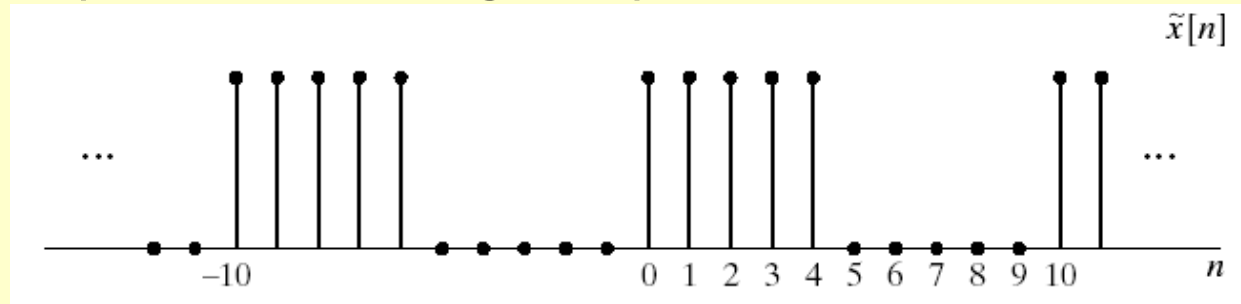
- We can represent the signal with the DFS coefficients as

$$\tilde{x}[n] = \sum_{r=-\infty}^{\infty} \delta[n - rN] = \frac{1}{N} \sum_{k=0}^{N-1} e^{j(2\pi/N)kn}$$



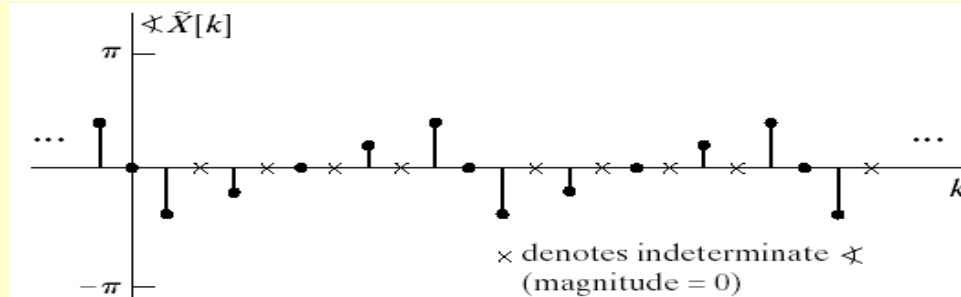
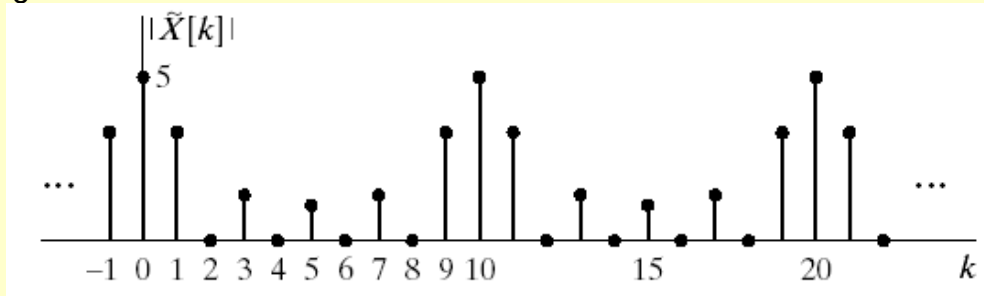
## Example 2

- DFS of an periodic rectangular pulse train



- The DFS coefficients

$$\tilde{X}[k] = \sum_{n=0}^4 e^{-j(2\pi/10)kn} = \frac{1 - e^{-j(2\pi/10)k5}}{1 - e^{-j(2\pi/10)k}} = e^{-j(4\pi k/10)} \frac{\sin(\pi k/2)}{\sin(\pi k/10)}$$





# Properties of DFS



- Linearity

$$\begin{array}{ccc} \tilde{x}_1[n] & \xleftrightarrow{\text{DFS}} & \tilde{X}_1[k] \\ \tilde{x}_2[n] & \xleftrightarrow{\text{DFS}} & \tilde{X}_2[k] \\ a\tilde{x}_1[n] + b\tilde{x}_2[n] & \xleftrightarrow{\text{DFS}} & a\tilde{X}_1[k] + b\tilde{X}_2[k] \end{array}$$

- Shift of a Sequence

$$\begin{array}{ccc} \tilde{x}[n] & \xleftrightarrow{\text{DFS}} & \tilde{X}[k] \\ \tilde{x}[n - m] & \xleftrightarrow{\text{DFS}} & e^{-j2\pi km/N} \tilde{X}[k] \\ e^{j2\pi nm/N} \tilde{x}[n] & \xleftrightarrow{\text{DFS}} & \tilde{X}[k - m] \end{array}$$

- Duality

$$\begin{array}{ccc} \tilde{x}[n] & \xleftrightarrow{\text{DFS}} & \tilde{X}[k] \\ \tilde{X}[n] & \xleftrightarrow{\text{DFS}} & N\tilde{x}[-k] \end{array}$$



# Symmetry Properties



Periodic Sequence (Period $N$ )	DFS Coefficients (Period $N$ )
1. $\tilde{x}[n]$	$\tilde{X}[k]$ periodic with period $N$
2. $\tilde{x}_1[n], \tilde{x}_2[n]$	$\tilde{X}_1[k], \tilde{X}_2[k]$ periodic with period $N$
3. $a\tilde{x}_1[n] + b\tilde{x}_2[n]$	$a\tilde{X}_1[k] + b\tilde{X}_2[k]$
4. $\tilde{X}[n]$	$N\tilde{x}[-k]$
5. $\tilde{x}[n - m]$	$W_N^{km} \tilde{X}[k]$
6. $W_N^{-\ell n} \tilde{x}[n]$	$\tilde{X}[k - \ell]$
7. $\sum_{m=0}^{N-1} \tilde{x}_1[m]\tilde{x}_2[n - m]$ (periodic convolution)	$\tilde{X}_1[k]\tilde{X}_2[k]$
8. $\tilde{x}_1[n]\tilde{x}_2[n]$	$\frac{1}{N} \sum_{\ell=0}^{N-1} \tilde{X}_1[\ell]\tilde{X}_2[k - \ell]$ (periodic convolution)
9. $\tilde{x}^*[n]$	$\tilde{X}^*[-k]$



# Symmetry Properties Cont'd



Periodic Sequence (Period $N$ )	DFS Coefficients (Period $N$ )
10. $\tilde{x}^*[-n]$	$\tilde{X}^*[k]$
11. $\mathcal{R}e\{\tilde{x}[n]\}$	$\tilde{X}_e[k] = \frac{1}{2}(\tilde{X}[k] + \tilde{X}^*[-k])$
12. $j\mathcal{I}m\{\tilde{x}[n]\}$	$\tilde{X}_o[k] = \frac{1}{2}(\tilde{X}[k] - \tilde{X}^*[-k])$
13. $\tilde{x}_e[n] = \frac{1}{2}(\tilde{x}[n] + \tilde{x}^*[-n])$	$\mathcal{R}e\{\tilde{X}[k]\}$
14. $\tilde{x}_o[n] = \frac{1}{2}(\tilde{x}[n] - \tilde{x}^*[-n])$	$j\mathcal{I}m\{\tilde{X}[k]\}$
Properties 15–17 apply only when $x[n]$ is real.	
15. Symmetry properties for $\tilde{x}[n]$ real.	$\left\{ \begin{array}{l} \tilde{X}[k] = \tilde{X}^*[-k] \\ \mathcal{R}e\{\tilde{X}[k]\} = \mathcal{R}e\{\tilde{X}[-k]\} \\ \mathcal{I}m\{\tilde{X}[k]\} = -\mathcal{I}m\{\tilde{X}[-k]\} \\  \tilde{X}[k]  =  \tilde{X}[-k]  \\ \angle\tilde{X}[k] = -\angle\tilde{X}[-k] \end{array} \right.$
16. $\tilde{x}_e[n] = \frac{1}{2}(\tilde{x}[n] + \tilde{x}[-n])$	$\mathcal{R}e\{\tilde{X}[k]\}$
17. $\tilde{x}_o[n] = \frac{1}{2}(\tilde{x}[n] - \tilde{x}[-n])$	$j\mathcal{I}m\{\tilde{X}[k]\}$





# Periodic Convolution



- Take two periodic sequences

$$\begin{aligned}\tilde{x}_1[n] &\xleftrightarrow{\text{DFS}} \tilde{X}_1[k] \\ \tilde{x}_2[n] &\xleftrightarrow{\text{DFS}} \tilde{X}_2[k]\end{aligned}$$

- Let's form the product

$$\tilde{X}_3[k] = \tilde{X}_1[k] \tilde{X}_2[k]$$

- The periodic sequence with given DFS can be written as

$$\tilde{x}_3[n] = \sum_{m=0}^{N-1} \tilde{x}_1[m] \tilde{x}_2[n-m]$$

- Periodic convolution is commutative

$$\tilde{x}_3[n] = \sum_{m=0}^{N-1} \tilde{x}_2[m] \tilde{x}_1[n-m]$$



## Periodic Convolution Cont'd



$$\tilde{x}_3[n] = \sum_{m=0}^{N-1} \tilde{x}_1[m] \tilde{x}_2[n-m]$$

- Substitute periodic convolution into the DFS equation

$$\tilde{X}_3[k] = \sum_{n=0}^{N-1} \left( \sum_{m=0}^{N-1} \tilde{x}_1[m] \tilde{x}_2[n-m] \right) W_N^{kn}$$

- Interchange summations

$$\tilde{X}_3[k] = \sum_{m=0}^{N-1} \tilde{x}_1[m] \left( \sum_{n=0}^{N-1} \tilde{x}_2[n-m] W_N^{kn} \right)$$

- The inner sum is the DFS of shifted sequence

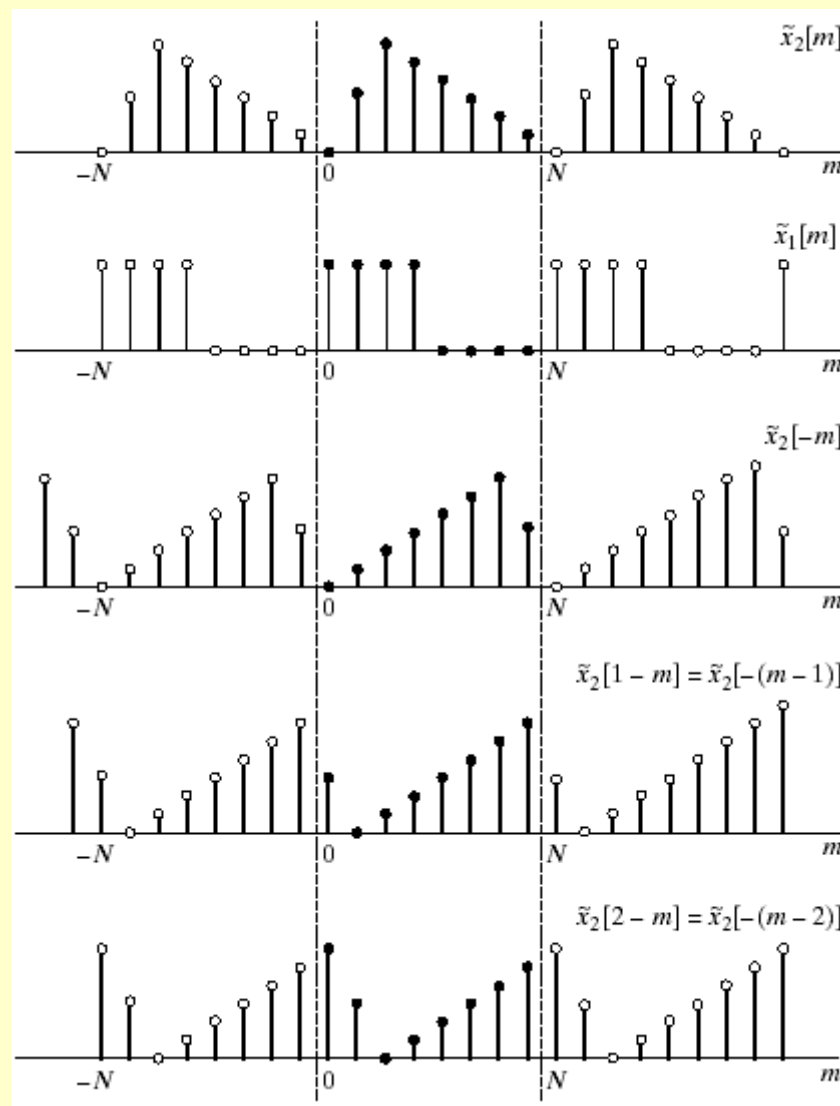
$$\sum_{n=0}^{N-1} \tilde{x}_2[n-m] W_N^{kn} = W_N^{km} \tilde{X}_2[k]$$

- Substituting

$$\tilde{X}_3[k] = \sum_{m=0}^{N-1} \tilde{x}_1[m] \left( \sum_{n=0}^{N-1} \tilde{x}_2[n-m] W_N^{kn} \right) = \sum_{m=0}^{N-1} \tilde{x}_1[m] W_N^{km} \tilde{X}_2[k] = \tilde{X}_1[k] \tilde{X}_2[k]$$



# Graphical Periodic Convolution





# The Fourier Transform of Periodic Signals



- Periodic sequences are not absolute or square summable
  - Hence they don't have a Fourier Transform
- We can represent them as sums of complex exponentials: DFS
- We can combine DFS and Fourier transform
- Fourier transform of periodic sequences
  - Periodic impulse train with values proportional to DFS coefficients

$$\tilde{X}(e^{j\omega}) = \sum_{k=-\infty}^{\infty} \frac{2\pi}{N} \tilde{X}[k] \delta\left(\omega - \frac{2\pi k}{N}\right)$$

- This is periodic with  $2\pi$  since DFS is periodic
- The inverse transform can be written as

$$\frac{1}{2\pi} \int_{0-\varepsilon}^{2\pi-\varepsilon} \tilde{X}(e^{j\omega}) e^{j\omega n} d\omega = \frac{1}{2\pi} \int_{0-\varepsilon}^{2\pi-\varepsilon} \sum_{k=-\infty}^{\infty} \frac{2\pi}{N} \tilde{X}[k] \delta\left(\omega - \frac{2\pi k}{N}\right) e^{j\omega n} d\omega$$

$$\frac{1}{N} \sum_{k=-\infty}^{\infty} \tilde{X}[k] \int_{0-\varepsilon}^{2\pi-\varepsilon} \delta\left(\omega - \frac{2\pi k}{N}\right) e^{j\omega n} d\omega = \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] e^{j\frac{2\pi k}{N}n}$$



## Example



- Consider the periodic impulse train

$$\tilde{p}[n] = \sum_{r=-\infty}^{\infty} \delta[n - rN]$$

- The DFS was calculated previously to be

$$\tilde{P}[k] = 1 \quad \text{for all } k$$

- Therefore the Fourier transform is

$$\tilde{P}(e^{j\omega}) = \sum_{k=-\infty}^{\infty} \frac{2\pi}{N} \delta\left(\omega - \frac{2\pi k}{N}\right)$$



## Relation between Finite-length and Periodic Signals



- Consider finite length signal  $x[n]$  spanning from 0 to  $N-1$
- Convolve with periodic impulse train

$$\tilde{x}[n] = x[n] * \tilde{p}[n] = x[n] * \sum_{r=-\infty}^{\infty} \delta[n - rN] = \sum_{r=-\infty}^{\infty} x[n - rN]$$

- The Fourier transform of the periodic sequence is

$$\tilde{X}(e^{j\omega}) = X(e^{j\omega}) \tilde{P}(e^{j\omega}) = X(e^{j\omega}) \sum_{k=-\infty}^{\infty} \frac{2\pi}{N} \delta\left(\omega - \frac{2\pi k}{N}\right)$$

$$\tilde{X}(e^{j\omega}) = \sum_{k=-\infty}^{\infty} \frac{2\pi}{N} X\left(e^{j\frac{2\pi k}{N}}\right) \delta\left(\omega - \frac{2\pi k}{N}\right)$$

- This implies that

$$\tilde{X}[k] = X\left(e^{j\frac{2\pi k}{N}}\right) = X(e^{j\omega}) \Big|_{\omega = \frac{2\pi k}{N}}$$

- DFS coefficients of a periodic signal can be thought as equally spaced samples of the Fourier transform of one period



# Example



- Consider the following sequence

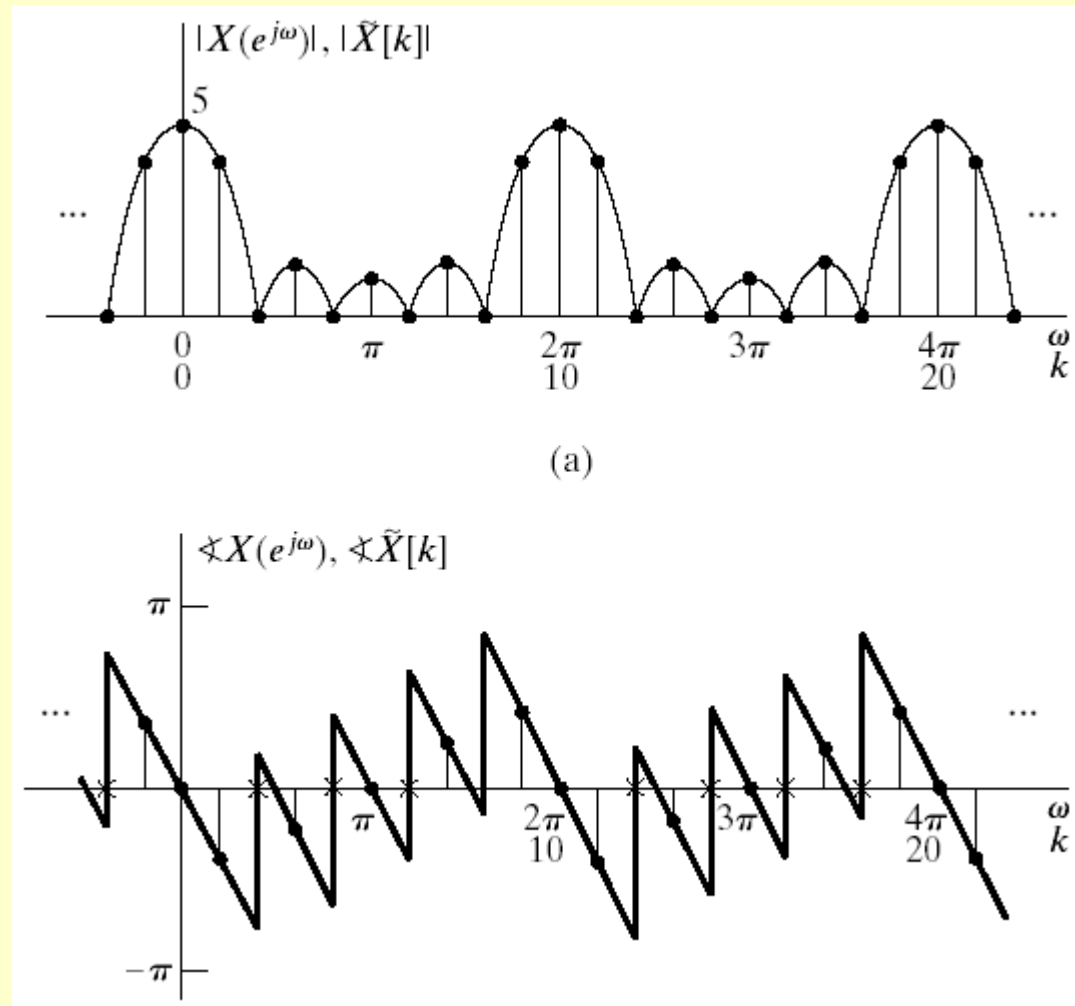
$$x[n] = \begin{cases} 1 & 0 \leq n \leq 4 \\ 0 & \text{else} \end{cases}$$

- The Fourier transform

$$X(e^{j\omega}) = e^{-j2\omega} \frac{\sin(5\omega/2)}{\sin(\omega/2)}$$

- The DFS coefficients

$$\tilde{X}[k] = e^{-j(4\pi k/10)} \frac{\sin(\pi k/2)}{\sin(\pi k/10)}$$





THANK YOU